A7330A

DC-DC CONVERTER BUCK (STEP-DOWN) 2.5A/40V INPUT,1MHZ SYNCHRONOUS STEP-DOWN CONVERTER

#### **DESCRIPTION**

The A7330A is a current mode monolithic buck switching regulator. Operating with an input range of 4.5V~40V, the A7330A delivers 2.5A of continuous output current with two integrated N-Channel MOSFETs. The internal synchronous power switches provide high efficiency without the use of an external Schottky diode. At light loads, the regulator operates in low frequency to maintain high efficiency and low output ripple. Current mode control provides tight load transient response and cycle-by-cycle current limit.

The A7330A guarantees robustness with short-circuit protection, thermal protection, current runaway protection, and input under voltage lockout.

The A7330A is available in PSOP8 package, which provides a compact solution with minimal external components.

#### ORDERING INFORMATION

Package Type	Part Number		
PSOP8	MP8	A7330AMP8R	
SPQ: 3,000psc/Reel	IVIP8	A7330AMP8VR	
Note	V: Halogen free Package R: Tape & Reel		
AiT provides all RoHS products			

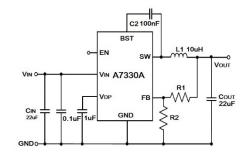
## **FEATURES**

- 4.5V to 40V Input Voltage Range
  2.5A output current
- High Efficiency: Up to 95%
- No Schottky Diode Required
- 0.8V Reference
- Slope Compensated Current Mode Control for Excellent Line and Load Transient Response
- Integrated internal compensation2
- Stable with Low ESR Ceramic Output Capacitors
- Input under voltage lockout
- Short circuit protection
- Thermal Shutdown
- Inrush Current Limit and Soft Start
- -40°C to +125°C Temperature Range

#### **APPLICATION**

- Distributed Power Systems
- Automotive Systems
- High Voltage Power Conversion
- Industrial Power Systems
- Battery Powered Systems

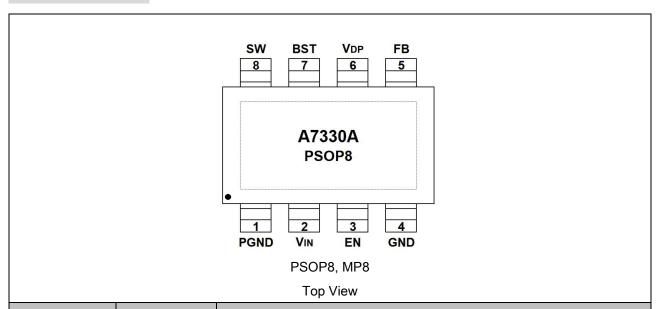
#### TYPICAL APPLICATION



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## PIN DESCRIPTION



Pin#	Symbol	Functions	
PSOP8	Symbol	Functions	
1	PGND	Power Ground	
2	VIN	Supply Voltage.	
3	EN	Enable pin for the IC. Drive this pin high to enable IC, low to disable.	
4	GND	Ground.	
5	Feedback Input. Connect an external resistor divider from the o		
3	ГВ	FB and GND to set V <sub>OUT</sub> .	
6	$V_{DP}$	Power for Power driver	
7	BST	Bootstrap pin. Connect a 100nF capacitor from this pin to SW.	
8 SW		Inductor Connection. Connect an inductor Between SW and the	
0	SW	regulator output.	
9	Thermal	Thermal PAD, must be connected to GND.	
9	PAD	Thermal FAD, must be connected to GND.	

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## ABSOLUTE MAXIMUM RATINGS

V <sub>IN</sub> , Input Voltage		$V_{SS}$ -0.3V ~ $V_{SS}$ +45V		
V <sub>SW</sub> , SW Voltage		$V_{SS}$ -0.3 $V \sim V_{IN}$ +0.3 $V$		
V <sub>BST</sub> , BST Voltage		$V_{SW}$ -0.3 $V \sim V_{SW}$ +6 $V_{SW}$		
V <sub>FB</sub> , FB Voltage		$V_{SS}$ -0.3 $V \sim V_{SS}$ +6 $V$		
V <sub>EN</sub> , EN Voltage		$V_{SS}$ -0.3 $V$ $\sim$ $V_{SS}$ +45 $V$		
P <sub>D</sub> , Power Dissipation PSOP8		600mW		
θ <sub>JC</sub> , Thermal Resistance		130°C/W		
θ <sub>JA</sub> , Thermal Resistance		170°C/W		
Topr, J Operating Ambient Temperature		-40°C ~ +125°C		
T <sub>STG</sub> , Storage Temperature		-40°C ~ +150°C		
ESD HBM (Human Body Mode)		2KV		

Stresses above may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated in the Electrical Characteristics are not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## **ELECTRICAL CHARACTERISTICS**

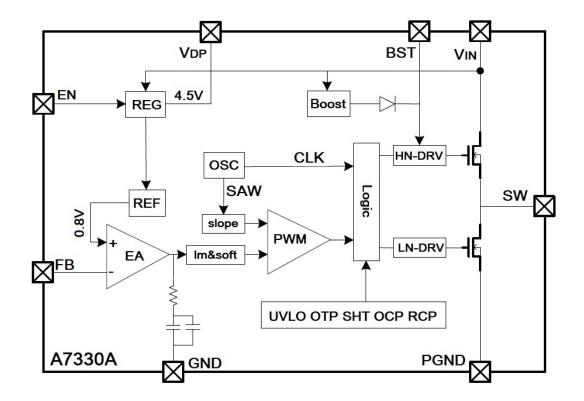
 $(V_{IN}=12V, V_{OUT}=5V, T_A=+25^{\circ}C, unless otherwise noted.)$ 

Parameter	Symbol	Conditions	Min.	Тур.	Max.	Unit
Input Voltage	VIN		4.5	-	40	V
V <sub>IN</sub> Under Voltage Protect	V <sub>UV</sub>		-	4.1	-	V
V <sub>IN</sub> Under Voltage Protect Hys.	V <sub>UV_HYS</sub>		-	0.3	-	V
Supply Current	I <sub>Q1</sub>	V <sub>EN</sub> =2V , V <sub>FB</sub> =1.2V	-	40	60	μA
Supply Shutdown Current	IsD		-1	-	1	μA
FB Voltage	V <sub>FB</sub>	T <sub>A</sub> =25°C , 4.5V≤V <sub>IN</sub> ≤40V	0.776	0.8	0.824	V
Switching Frequency	Fosc		-	1	-	MHz
Minimum on Time	T <sub>ON_MIN</sub>		-	200	-	ns
Maximum Duty Cycle	D <sub>MAX</sub>		-	95	-	%
High Side Switch on Resistance	R <sub>DSON_H</sub>		-	200	-	mΩ
Low Side Switch on Resistance	R <sub>DSON_L</sub>		-	120	-	mΩ
High Side Current Limit	I <sub>LIM</sub>		-	3.5	-	Α
EN Rising Threshold	V <sub>ENH</sub>		-	1.3	-	V
EN Falling Threshold	V <sub>ENL</sub>		-	1.2	-	V
EN Input Current	I <sub>EN</sub>	V <sub>EN</sub> =5V	-	1	-	μA
Thermal Shutdown	T <sub>SHD</sub>		-	155	-	°C
Thermal Shutdown Hys.	T <sub>SHD_HYS</sub>		-	25	-	°C

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## **BLOCK DIAGRAM**



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## **DETAILED INFORMATION**

The A7330A is a synchronous, current-mode, step-down regulator. It regulates input voltages from 4.5V to 40V, and is capable of supplying up to 2.5A of load current.

#### **Current-Mode Control**

The A7330A utilizes current-mode control to regulate the output voltage. The output voltage is measured at the FB pin through a resistive voltage divider and the error is amplified by the internal trans-conductance error amplifier. Output of the internal error amplifier is compared with the switch current measured internally to control the output current.

#### **PFM Mode**

The A7330A operates in PFM mode at light load. In PFM mode, switch frequency decreases when load current drops to boost power efficiency at light load by reducing switch-loss, while switch frequency increases when load current rises, minimizing output voltage ripples.

#### Shutdown Mode

The A7330A shuts down when voltage at EN pin is below 0.3V. The entire regulator is off and the supply current consumed by the A7330A drops below 0.1uA.

#### **Power Switch**

N-Channel MOSFET switches are integrated on the A7330A to down convert the input voltage to the regulated output voltage. Since the top MOSFET needs a gate voltage great than the input voltage, a boost capacitor connected between BST and SW pins is required to drive the gate of the top switch. The boost capacitor is charged by the internal 4.5V rail when SW is low.

#### V<sub>IN</sub> Under Voltage Protection

A resistive divider can be connected between  $V_{IN}$  and ground, with the central tap connected to EN, so that when  $V_{IN}$  drops to the pre-set value, EN drops below 1.2V to trigger input under voltage lockout protection.

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#### **Over-Current-Protection and Hiccup**

At start-up, due to the high voltage at input and low voltage at output, current inertia of the output inductance can be easily built up, resulting in a large start-up output current. In the A7330A used current limit of low side power mosfet to control the output current at start-up.

#### **Output Short Protection**

When output is shorted to ground, output current rapidly reaches its peak current limit and the top power switch is turned off. And the bottom power switch is turned on and stay on until the output current falls below the current limit. When output current is below the current limit, the top power switch will be turned on again and if the output short is still present, the top power switch is turned off when the peak current limit is reached and the bottom power switch is turned on. This cycle goes on until the output short is removed and the regulator comes into normal operation again.

#### Thermal Protection

When the temperature of the A7330A rises above 160°C, it is forced into thermal shut-down. Only when core temperature drops below 140°C can the regulator become active again.

#### **Setting the Output Voltage**

The external resistor divider is used to set the output voltage. Choose R1 and R2 follow the next table or calculated by following equation, where the internal reference voltage  $V_{REF}$ =0.8V.

$$\frac{R1}{R2} = \frac{V_{OUT} - V_{REF}}{V_{REF}}$$

Vout(V)	R2(kΩ)	R1(kΩ)
2.5	24	51
3.3	24	75
5.0	24	126

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#### **Inductor Selection**

A  $4.7\mu H$  to  $22\mu H$  inductor with a DC current rating of at least 25% percent higher than the maximum load current is recommended for most applications. For highest efficiency, the inductor DC resistance should be less than  $15m\Omega$ . For most designs, the inductance value can be derived from the following equation.

$$L > \frac{V_{OUT} X(V_{IN} - V_{OUT})}{V_{IN} X \Delta I_L X f_{OSC}}$$
 Where  $\Delta I_L$  is the inductor ripple current.

Choose ∆I<sub>L</sub> to be approximately 30% of the maximum load current.

Under light load conditions below 100mA, larger inductance is recommended for improved efficiency.

Note: If the output voltage is less than 2.5V, it is recommended to use a larger inductance and add a small capacitance (10nF) parallel to R1.

#### Input Capacitor Selection

The input capacitor reduces the surge current drawn from the input and switching noise from the device. The input capacitor impedance at the switching frequency should be less than input source impedance to prevent high frequency switching current passing to the input. A low ESR input capacitor sized for maximum RMS current must be used. Ceramic capacitors with X5R or X7R dielectrics are highly recommended because of their low ESR and small temperature coefficients. A 22µF electrolytic capacitor for most applications is sufficient. Include a capacitor with a value of 0.1µF for high-frequency filtering and place it as close as possible to the device pins.

#### **Output Capacitor Selection**

The output capacitor ( $C_{OUT}$ ) is required to maintain the DC output voltage, and the capacitance value determines the output ripple voltage. The output voltage ripple can be calculated by:

$$\Delta V_{OUT} = \frac{V_{OUT} \times (V_{IN} - V_{OUT})}{V_{IN} \times f_{OSC} \times L} \times (R_{ESR} + \frac{1}{8 \times f_{OSC} \times C_{OUT}})$$

Where L is the inductor value and  $R_{\text{ESR}}$  is the equivalent series resistance (ESR) value of the output capacitor. The output capacitor can be low ESR electrolytic, tantalum or ceramic, which lower ESR capacitors get lower output ripple voltage.

The characteristics of the output capacitor also affect the stability of the regulation system, and a 10uF ceramic capacitor is recommended in typical application. The A7330A can be optimized for a wide range of capacitance and ESR values.

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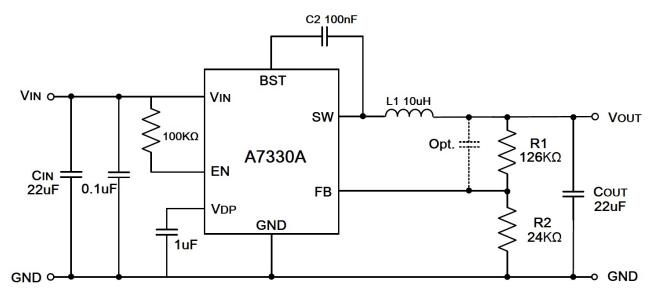


#### **Application Information**

For minimum noise problem and best operating performance, the PCB is preferred to following the guidelines as reference.

- 1. Place the input decoupling capacitor as close to A7330A (V<sub>IN</sub> pin and GND) as possible to eliminate noise at the input pin. The loop area formed by input capacitor and GND must be minimized.
- 2. Put the feedback trace as far away from the inductor and noisy power traces as possible.
- 3. The ground plane on the PCB should be as large as possible for better heat dissipation.

## Reference Design



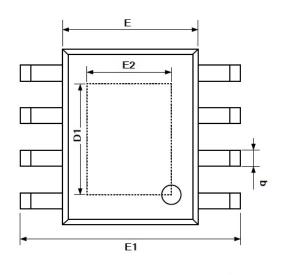
 $V_{IN}:6.5V\sim40V, V_{OUT}:5.0V, I_{OUT}:0\sim2.5A$ 

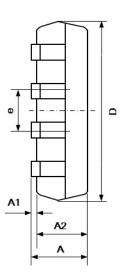
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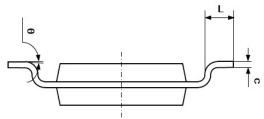
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## PACKAGE INFORMATION

Dimension in PSOP8 (Unit: mm)







C) mah al	Millimeters			
Symbol	Min	Max		
Α	1.350	1.750		
A1	0.050	0.150		
A2	1.350	0.550		
b	0.330	0.510		
С	0.170	0.250		
D	4.700	5.100		
D1	3.202	3.420		
Е	3.800	4.000		
E1	5.800	6.200		
E2	2.313	2.513		
е	1.270 BSC			
L	0.400	1.270		
θ	0°	8°		

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